



The University of Sheffield

*Department of Automatic Control & Systems  
Engineering*

# *Linearization Methods and Control of Nonlinear Systems – Two Cases*

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# *Importance of studying Nonlinear Systems*

- No model of a real system is truly linear even if it is profitable to study their linear approximation.
- Nonlinear systems may show complex effects (chaos, bifurcations, etc) that cannot be anticipated.
- It brings many disciplines together: mathematics, physics, biology, chemistry, engineering, economics, medicine, etc.

# *Importance of Linearization Methods*

**“They allow to adopt linear control techniques to analyze nonlinear problems”**

# *Some Linearization Techniques*

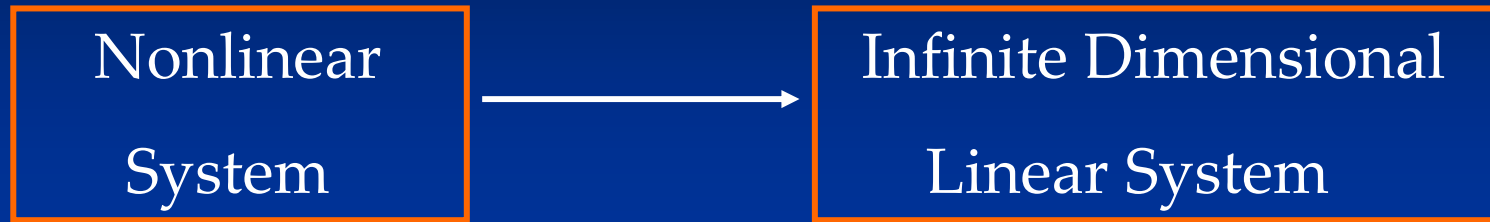
- Jacobian Linearization
- **Carleman Linearization**
- Linearization using Lie Series
- **Iteration Technique**
- Feedback linearization
- Linearization via changes of variables

# *Carleman Linearization and Lyapunov Stability Theory*

Navarro Hernandez, C. and Banks, S.P. (2004), A Generalization of Lyapunov's Equation to Nonlinear Systems, *NOLCOS 2004*, Stuttgart, Germany, September 2004

Banks, S.P. and Navarro Hernandez, C. (2005), A New Proof of McCann's Theorem and the Generalization of Lyapunov's Equation to Nonlinear Systems, *IJICIC*, to appear.

# Carleman Linearization



- Kowalski, K. and Steeb, W. *Nonlinear dynamical systems and Carleman linearization*. World Scientific Publishing Co. Singapore, 1991.
- Gaude, Brian. *Solving nonlinear aeronautical problems using the Carleman Linearization Method*. Sandia National Laboratories Report, U.S.A, 2001.

# *Applications to Control*

- **Domains of attraction of nonlinear feedback systems** (Loparo K.A. and Blankenship (1978) Estimating the domain of attraction of nonlinear feedback systems. *IEEE Trans. Aut. Control*, AC-23, 602)
- **Design of observers with linear error dynamics** (Deutscher, J. (2003) Asymptotically exact input-output linearization using Carleman linearization. *ECC2003 Cambridge, UK*)
- **Solutions of Lotka-Volterra models** (Steeb and Wilhelm, 1980)
- **Power series expansions for nonlinear systems** (Brenig and Fairén, 1981)
- **Construction of approximate Monte-Carlo-like solutions to nonlinear integral equations** (Ermakov, 1984)
- **Study of nonlinear partial differential equations** (Kowalski, 1988)

# Carleman Linearization

Consider the following nonlinear system

$$\dot{x} = f(x)$$

the Taylor series can be written as

$$f(x) = \sum_{\ell} A^{\ell} x^{[\ell]}$$

where

$$A^{\ell} = \left\{ \frac{1}{\ell!} \frac{\partial^{\ell} f(0)}{\partial x_{i_1} \cdots \partial x_{i_{\ell}}} \right\} \quad x^{[\ell]} = \underbrace{x \otimes x \otimes \dots \otimes x}_{\ell\text{-times}}$$



To obtain the linear system

$$x^{[k]} = \underbrace{x \otimes x \otimes \dots \otimes x}_{k\text{-times}}$$

$$\dot{x}^{[k]} = \sum_{i=1}^k x \otimes \dots \otimes \underbrace{\dot{x}}_i \otimes \dots \otimes x$$

$$= \sum_{i=1}^k x \otimes \dots \otimes f(x) \otimes \dots \otimes x$$

$$\dot{x}^{[k]} = \sum_{i=1}^k x \otimes \dots \otimes \sum_{\ell} A^{\ell} x^{[\ell]} \otimes \dots \otimes x$$

Therefore

$$\dot{x}^{[k]} = \sum_{\ell} A_k^{\ell} x^{[k+\ell-1]}$$

$$A_k^{\ell} = \sum_{i=1}^k I \otimes \dots \otimes I \otimes A^{\ell} \otimes I \otimes \dots \otimes I$$

$$A_k^{\ell} = A_1^{\ell} \otimes I^{[k-1]} + I \otimes A_{k-1}^{\ell}$$

Setting

$$w = (x, x \otimes x, x \otimes x \otimes x, \dots)^T$$

We obtain the infinite linear system

$$\dot{w} = Aw$$

$$A = \begin{pmatrix} A_1^1 & A_1^2 & A_1^3 & \dots & \dots & \dots \\ 0 & A_2^1 & A_2^2 & A_2^3 & \dots & \dots \\ 0 & 0 & A_3^1 & A_3^2 & A_3^3 & \dots \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \end{pmatrix}$$

# *Lyapunov's Stability*

$V(x)$  is a Lyapunov function for the equilibrium  $x^*$   
if

1.  $V(x) > 0$

2.  $\dot{V}(x) < 0$

$$\forall x \neq x^*$$

Problem  $\longrightarrow$  Find the Lyapunov function

# *Lyapunov's Equation*

Linear time invariant system

$$\dot{x} = Ax$$

is asymptotically stable iff

$$A^T P + PA = -Q$$

The Lyapunov function is

$$V = x^T P x$$

- Vanelli, A. and Vidyasagar, M. (1985). Maximal Lyapunov functions and domains of attraction for autonomous nonlinear systems, *Automatica*, **21**, 69-80. (Lyapunov functions that are rational rather than polynomial)
- Camilli, F., Grüne L. and Wirth, F. Zubov's method for perturbed differential equations. NOLCOS 2004. Stuttgart, Germany (Generation of Lyapunov functions and domains of attraction using Zubov method)

# *Lyapunov's Equation and McCann's Theorem*

Consider the linear system  $\dot{w} = Aw$

and let  $P$  satisfy the Lyapunov equation

$$A^T P + PA = -I$$

Then,

$$V = \langle (x, x \otimes x, x \otimes x \otimes x, \dots), P(x, x \otimes x, x \otimes x \otimes x, \dots) \rangle$$

$$\dot{V} = -\|(x, x \otimes x, x \otimes x \otimes x, \dots)\|^2 = -(e^{\|x\|^2} - 1)$$

*The nonlinear system*

$$\dot{x} = f(x)$$

*is globally asymptotically stable iff*

$$A^T P + P A = -I \quad (*)$$

*is soluble for  $P$  (positive definite) .*

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*McCann's Theorem* - Global asymptotically stable dynamical systems are equivalent to linear systems.  
(McCann, 1979)

*If  $\dot{x} = f(x)$  is globally asymptotically stable there is a positive definite solution of (\*)*



# Example #1

Consider the nonlinear system  $\dot{x} = -x + x^3$

and obtain the infinite linear system  $\dot{w} = Aw$

$$A = \begin{pmatrix} -1 & 0 & 1 & 0 & \dots & \dots & \dots \\ 0 & -2 & 0 & 2 & 0 & \dots & \dots \\ 0 & 0 & -3 & 0 & 3 & 0 & \dots \\ \dots & \dots & \dots & \dots & \dots & \dots & \dots \end{pmatrix}$$

$$w = (x, x^2, x^3, x^4, \dots)^T$$

## Truncate $A$ and solve Lyapunov Equation

$$\begin{pmatrix} -1 & 0 & 0 \\ 0 & -2 & 0 \\ 1 & 0 & -3 \end{pmatrix} \mathbf{P} + \mathbf{P} \begin{pmatrix} -1 & 0 & 1 \\ 0 & -2 & 0 \\ 0 & 0 & -3 \end{pmatrix} = -\mathbf{I}$$

## Obtain Lyapunov function

$$V = \frac{1}{2} x^2 + \frac{1}{2} x^4 + \frac{5}{24} x^6$$

$$\dot{V} = (x^2 - 1)(x^2 + 2x^4 + \frac{5}{4} x^6)$$

## Truncating $A$ to a 7x 7 matrix

$$V = \frac{1}{2} x^2 + \frac{1}{2} x^4 + \frac{1}{2} x^6 + \frac{1}{2} x^8 + \frac{253}{640} x^{10} + \frac{1087}{3840} x^{12} + \frac{1343}{10752} x^{14}$$

$$\dot{V} = (x^2 - 1)(x^2 + \frac{7680}{3840} x^4 + \frac{11520}{3840} x^6 + \frac{15360}{3840} x^8 + \frac{15180}{3840} x^{10} + \frac{13044}{3840} x^{12} + \frac{6715}{3840} x^{14})$$

Fig. 1  $V(x)$  vs  $x$

$$V = \frac{1}{2}x^2 + \frac{1}{2}x^4 + \frac{5}{24}x^6$$

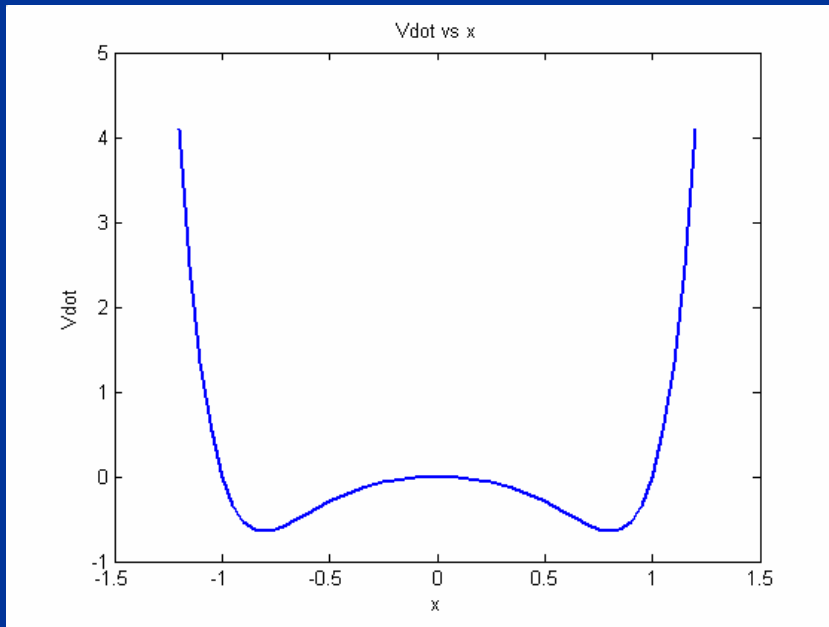
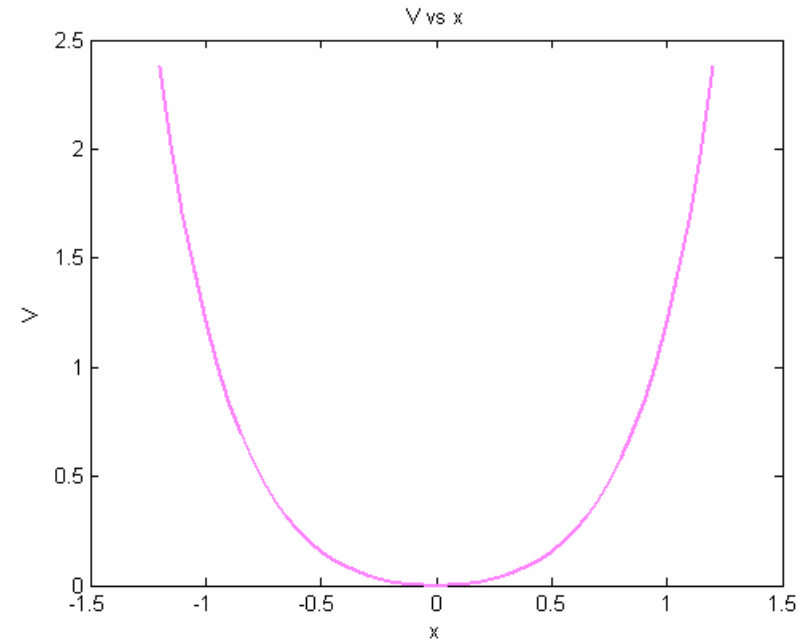


Fig. 2  $\dot{V}(x)$  vs  $x$

$$\dot{V} = (x^2 - 1)(x^2 + 2x^4 + \frac{5}{4}x^6)$$

# Example #2

Consider the nonlinear system

$$\dot{x} = -x + 4y$$

$$\dot{y} = -x - y^3$$

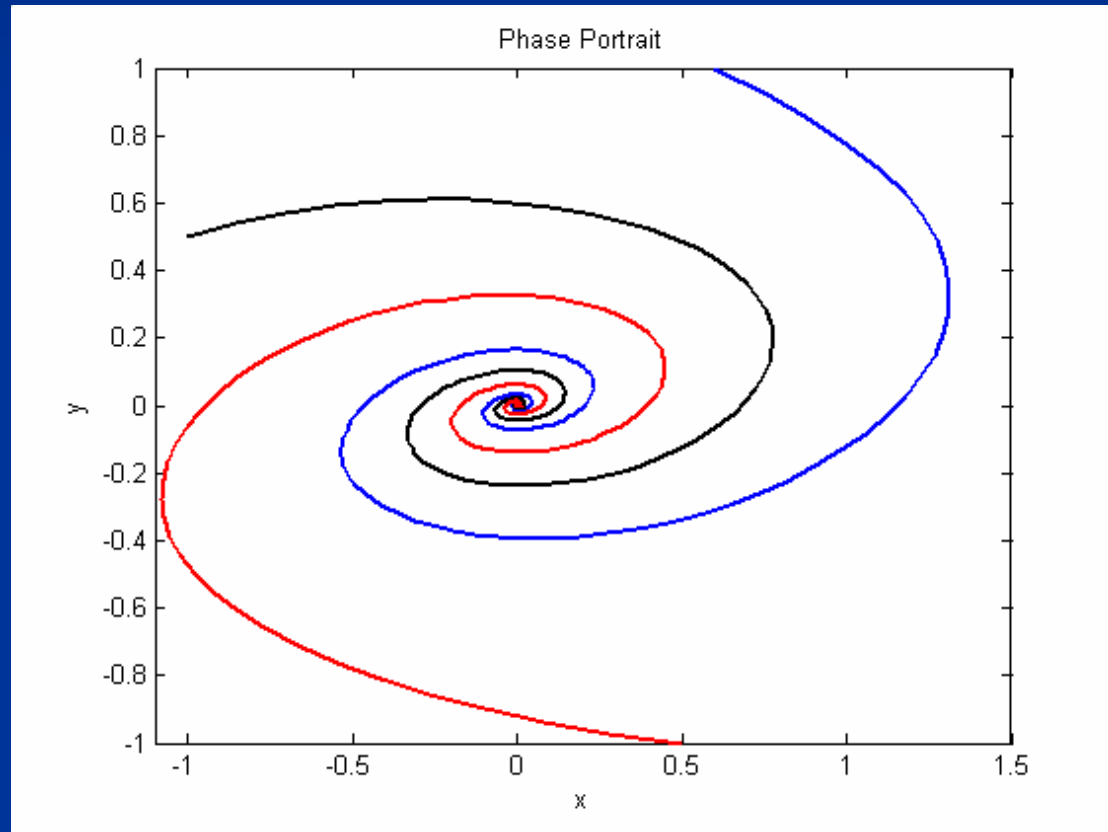


Fig. 3 Phase Portrait

Nonlinear System:  $\dot{x} = -x + 4y$   
 $\dot{y} = -x - y^3$

Now

$$w = (\underbrace{x, y}_{\mathbf{x}}, \underbrace{x^2, xy, yx, y^2}_{\mathbf{x} \otimes \mathbf{x}}, \underbrace{x^3, x^2 y, x^2 y, xy^2, x^2 y, xy^2, xy^2, y^3, \dots}_{\mathbf{x} \otimes \mathbf{x} \otimes \mathbf{x}})^T$$

$$A = \begin{pmatrix} A_1^1 & A_1^2 & A_1^3 & 0 & 0 & 0 \\ 0 & A_2^1 & A_2^2 & A_2^3 & 0 & 0 \\ 0 & 0 & A_3^1 & A_3^2 & A_3^3 & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots & \vdots \end{pmatrix}$$

$$A_1^1 = \begin{pmatrix} -1 & 4 \\ -1 & 0 \end{pmatrix} \quad A_1^2 = \begin{pmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{pmatrix} \quad A_1^3 = \begin{pmatrix} 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & -1 \end{pmatrix}$$

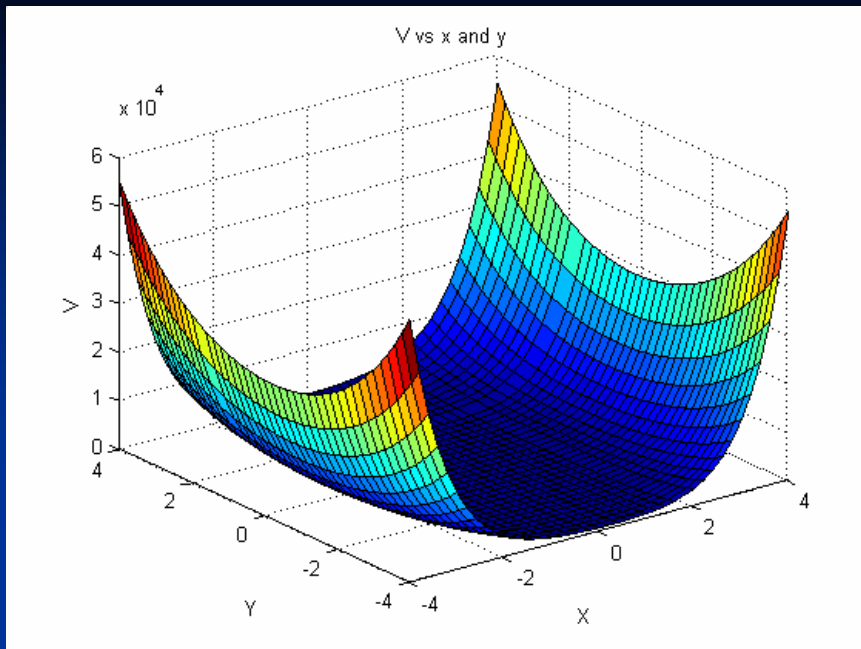
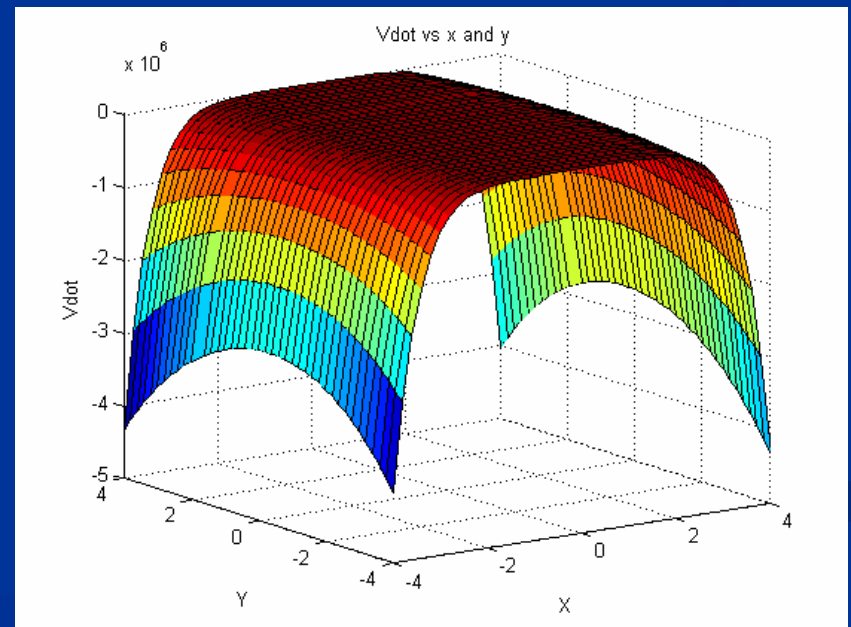


Fig. 4  $V(x,y)$  vs  $x,y$

Fig. 5  $\dot{V}(x,y)$  vs  $x,y$



# *Conclusions*

- A method to find a Lyapunov function for stable nonlinear systems by using a defined Lyapunov equation.
- By increasing the number of terms of the truncation of the infinite-dimensional Lyapunov equation it is expected that the method will approximate in a better way the basin of attraction of the systems.
- It is computationally difficult as the operator  $A$  grows exponentially and the Lyapunov expressions are complicated

# *Future Work*

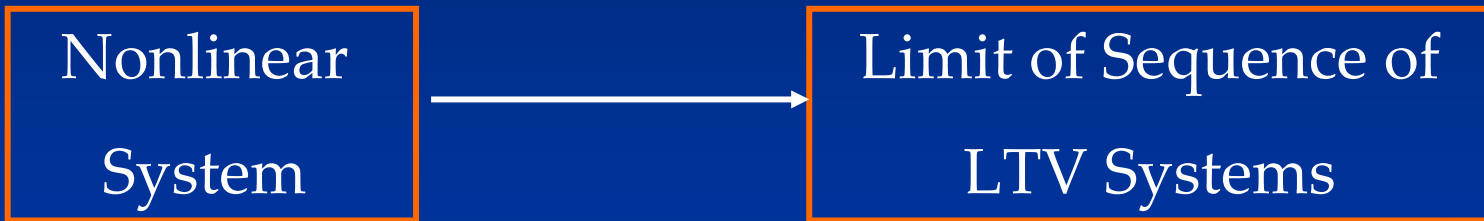
- Study of the required level of truncation of the Lyapunov equation
- Use of the Lyapunov functions for Control Design by introducing unknown parameters in the operator  $\mathbf{A}$
- Definition of the Lyapunov function by expanding the nonlinear system using orthogonal functions



# *Iteration Technique and Fault Detection for Nonlinear Systems*

Navarro Hernandez, C, Crusca, F., Aldeen, M. and Banks, S. P. (2005), *Fault Detection for Nonlinear Systems using Linear Approximations, To be presented at IFAC2005, Prague, Cz. July 2005*

# Iteration Technique



- Tomas-Rodriguez, M., Banks, S., (2003) Linear approximations to nonlinear dynamical systems with applications to stability and spectral theory, *IMA Journal of Mathematical Control and Information*, 20, 89-103.

# *Applications to Control*

- **Stability and spectral theory**

Tomas-Rodriguez M. and Banks, S., (2003) Linear approximations to nonlinear dynamical systems with applications to stability and spectral theory, *IMA Journal of Mathematical Control and Information*, **20**, 89-103.

- **Design of Observers**

Navarro Hernandez, C., Banks, S.P. and Aldeen, M. (2003) Observer design for nonlinear systems using linear approximations, *IMA Journal of Mathematical Control and Information*, **20**, 359-370.

- **Pole Placement for Nonlinear Systems**

Tomas-Rodriguez M. and Banks, S.P. (2004). Pole placement for nonlinear systems., *NOLCOS 2004*, Stuttgart, Germany.

- **Optimal Control**

Çimen, T and Banks, S.P. (2004). Nonlinear optimal tracking control with application to super-tankers for autopilot design, *Automatica*.

Having the nonlinear system

$$\dot{x} = A(x)x, x(0) = x_0 \in \mathfrak{R}^n.$$

and introducing the sequence of linear time varying equations:

$$\dot{x}^{[1]}(t) = A(x_0)x^{[1]}(t), x^{[1]}(0) = x_0$$

$\vdots$

$$\dot{x}^{[i]}(t) = A(x^{[i-1]}(t))x^{[i]}(t), x^{[i]}(0) = x_0$$

where  $i$ =number of approximations, it can be shown that the solution of this sequence converges to the solution of the original nonlinear system if the Lipschitz condition is satisfied.

$$\text{Lipschitz condition} \quad \|A(x) - A(y)\| \leq \alpha \|x - y\|$$

# Solution to Van der Pol oscillator

$$\begin{pmatrix} \dot{x}_1 \\ \dot{x}_2 \end{pmatrix} = \begin{pmatrix} -x_1^2 + 1 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}$$

and for the  $i$ th approximation,

$$\begin{pmatrix} \dot{x}_1^{[i]}(t) \\ \dot{x}_2^{[i]}(t) \end{pmatrix} = \begin{pmatrix} -(x_1^{[i-1]}(t))^2 + 1 & 1 \\ -1 & 0 \end{pmatrix} \begin{pmatrix} x_1^{[i]}(t) \\ x_2^{[i]}(t) \end{pmatrix}$$

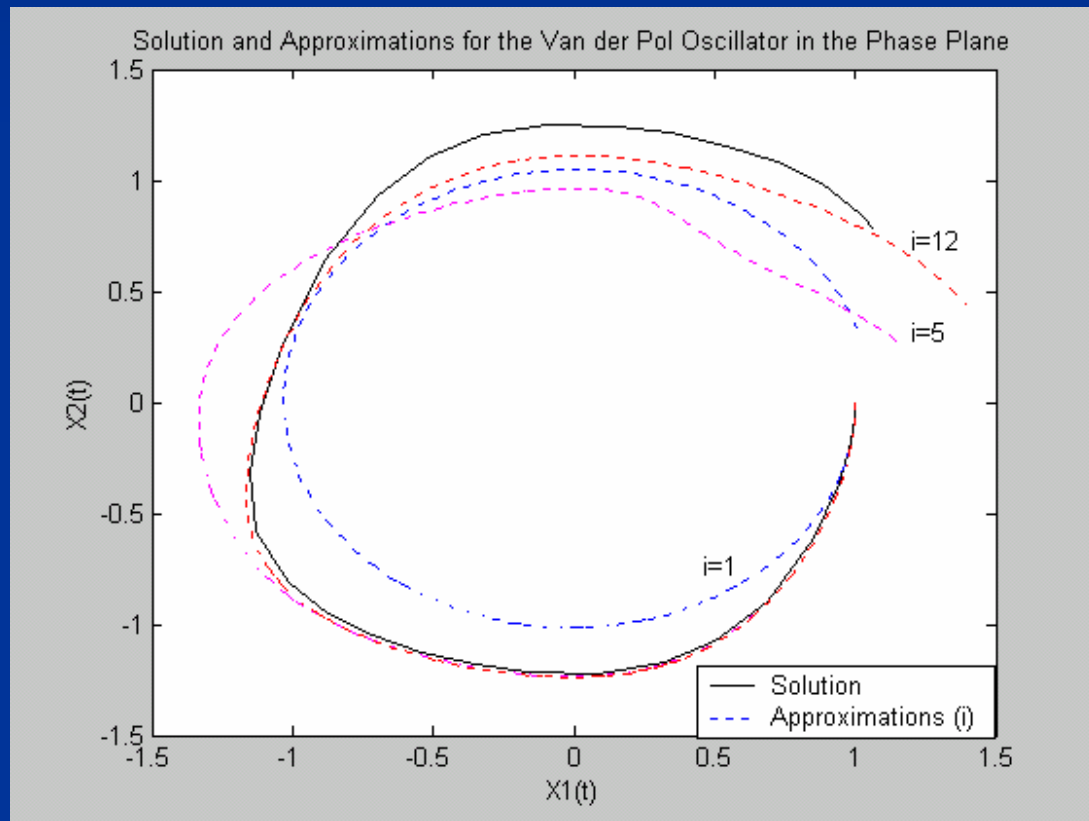


Fig 6. Solution and Approximations of the Van der Pol Oscillator

# *Fault Detection for Nonlinear Systems*

**Fault** – Unacceptable change or failure of at least one parameter or system characteristic from its designed or normal operating conditions

**Fault Detection** - Determination of system faults and the time of the on-set faults

**Model based-approach**  
(based on the mathematical  
model)

State Estimation-Based Approach

Reconstruct plant states and generate residuals  
by comparing estimated outputs with the  
measurements

Type of Fault: Abrupt and Additive

# Fault Detection for LTV Systems

LTV  
SYSTEM:

$$\dot{x}(t) = Ax + B_u u + B_w w + F_1 \mu_1 + F_2 \mu_2$$
$$y(t) = Cx + v, x(t_0) = x_0$$

Where  $u$  control input  $v$  sensor noise  
 $y$  measurement  $\mu_1, \mu_2$  faults (functions of time)  
 $w$  process noise  $F_1, F_2$  faults directions

Find linear observer and residual

$$\dot{\hat{x}}(t) = A\hat{x} + B_u u + L(y - C\hat{x})$$
$$r = \hat{H}(y - C\hat{x})$$

**Objective**

To find a residual primarily affected by the target fault and minimally by the nuisance faults



## *LTV Designs*

- Chen,R., Mingori,D. and Speyer, J. (2003). Optimal stochastic fault detection filter, *Automatica*, 39, 377-390.
- Xu,A., Zhang, Q. (2002). Fault Detection and Isolation based on Adaptive Observers for Linear Time Varying Systems , *15th Triennial World Congress*, Barcelona, Spain.
- Edelmayer,A., Bokor,J., Szigeti,F. and Keviczky, L. (1997). Robust Detection Filter in the Presence of Time-Varying System Perturbations, *Automatica*, 33, No. 3, 471-475.

*Design proposed:* “Optimal stochastic fault detection filter”  
(Chen, Mingori and Speyer)

- ❖ Algorithm is easy to program
- ❖ Extends the result of the UIO to the time varying case

## Steps in design:

1. Find the filter gain by solving a Riccati equation

$$\dot{P} = AP + PA^T - PC^T V^{-1} CP + \frac{1}{\gamma} F_2 Q_2 F_2^T - F_1 Q_1 F_1^T + B_w Q_w B_w^T, P(t_0) = P_0$$

$$L = PC^T V^{-1}$$

2. Find the projector

$$\hat{H} = [\rho_m \rho_{m-1} \dots \rho_{p_2+1}] [\rho_m \rho_{m-1} \dots \rho_{p_2+1}]^T$$

where  $CPC^T \rho = \rho \lambda$ ,  $p_2 = \dim F_2$ ,  $\lambda_1 \geq \lambda_2 \geq \dots \geq \lambda_m$

3. Calculate the estimate

$$\dot{\hat{x}}(t) = A\hat{x} + B_u u + L(y - C\hat{x})$$

4. Obtain the residual

$$r = \hat{H}(y - C\hat{x})$$

- Alcorta Garcia, E. and Frank, P.M. (1997). Deterministic nonlinear observer-based approaches to fault diagnosis: a survey, *Control Engineering Practice*, 5, 663-670.
- Yang, H. and Saif, S.M. (1996) Monitoring and diagnostics of a class of nonlinear system using a nonlinear unknown input observer, *Proceedings of the 1996 IEEE International Conference on Control Applications*, Dearborn, MI, September 15-18. (Decomposition of the nonlinear system)
- Frank, P.M., (1994). On-line fault detection in uncertain nonlinear systems using diagnostic observers: a survey, *Int. J. Systems Sci.* 25, 2129-2154.
- Seliger, R. and Frank, P.M. (1991) Fault-Diagnosis by disturbance decoupled nonlinear observers, *Proceedings of the 30th IEEE Conference on Decision and Control*, Brighton, England, 2248-2253. (Nonlinear state transformations)

# *Fault Detection for Nonlinear Systems*

Given a nonlinear system  $\dot{x}(t) = A(x)x(t) + B(x)u(t) + F_1(x)\mu_1(t) + F_2(x)\mu_2(t)$   
 $y(t) = C(x)x(t), x(t_0) = x_0$

## *1. Reduction to a sequence of linear time varying approximations*

$$\dot{x}^{[i]}(t) = A(x^{[i-1]}(t))x^{[i]}(t) + B(x^{[i-1]}(t))u^{[i]}(t) + F_1(x^{[i-1]}(t))\mu_1(t) + F_2(x^{[i-1]}(t))\mu_2(t)$$
$$y^{[i]}(t) = C(x^{[i-1]}(t))x^{[i]}(t), x^{[i]}(0) = x_0$$

## *2. Design of the Fault Detection Filter at each TV approximation*

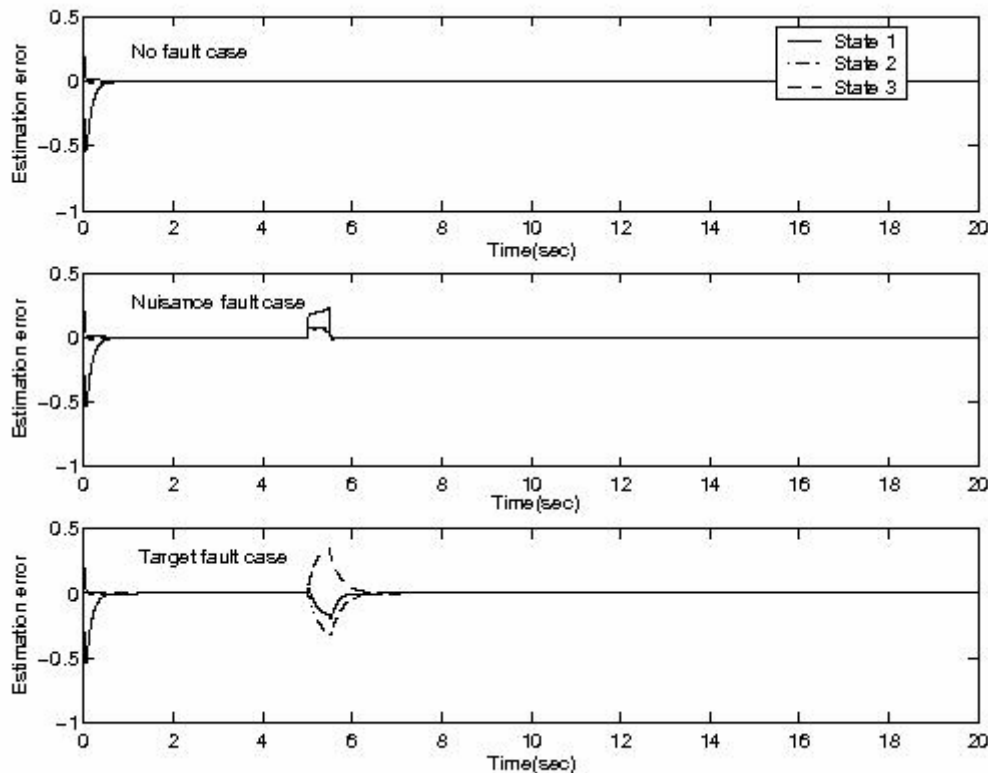
$$\dot{\hat{x}}^{[i]}(t) = A(\hat{x}^{[i-1]}(t))\hat{x}^{[i]}(t) + B(x^{[i-1]}(t))u^{[i]}(t) + L^{[i]}(t)(y^{[i]}(t) - C(\hat{x}^{[i-1]}(t))\hat{x}^{[i]}(t))$$
$$r^{[i]}(t) = \hat{H}^{[i]}(t)(y^{[i]}(t) - C(x^{[i-1]}(t))\hat{x}^{[i]}(t))$$

## *3. Obtain the residual at the final approximation*

# Nonlinear Example

$$\dot{x}(t) = \begin{pmatrix} -8 & x_2 & 0 \\ x_3 & -5 & x_1 \\ -x_2 & x_3 & -x_2 \end{pmatrix} x(t) + \begin{pmatrix} 1 \\ 0 \\ 1 \end{pmatrix} \mu_1(t) + \begin{pmatrix} 5 - 2 \cos(x_1) \\ 1 \\ 1 + \sin(x_2) \end{pmatrix} \mu_2(t)$$

$$y(t) = \begin{pmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \end{pmatrix} x(t)$$



$$\mu_1 = \mu_2$$

impulses of magnitude 3 at t=5 to t=5.5 sec.

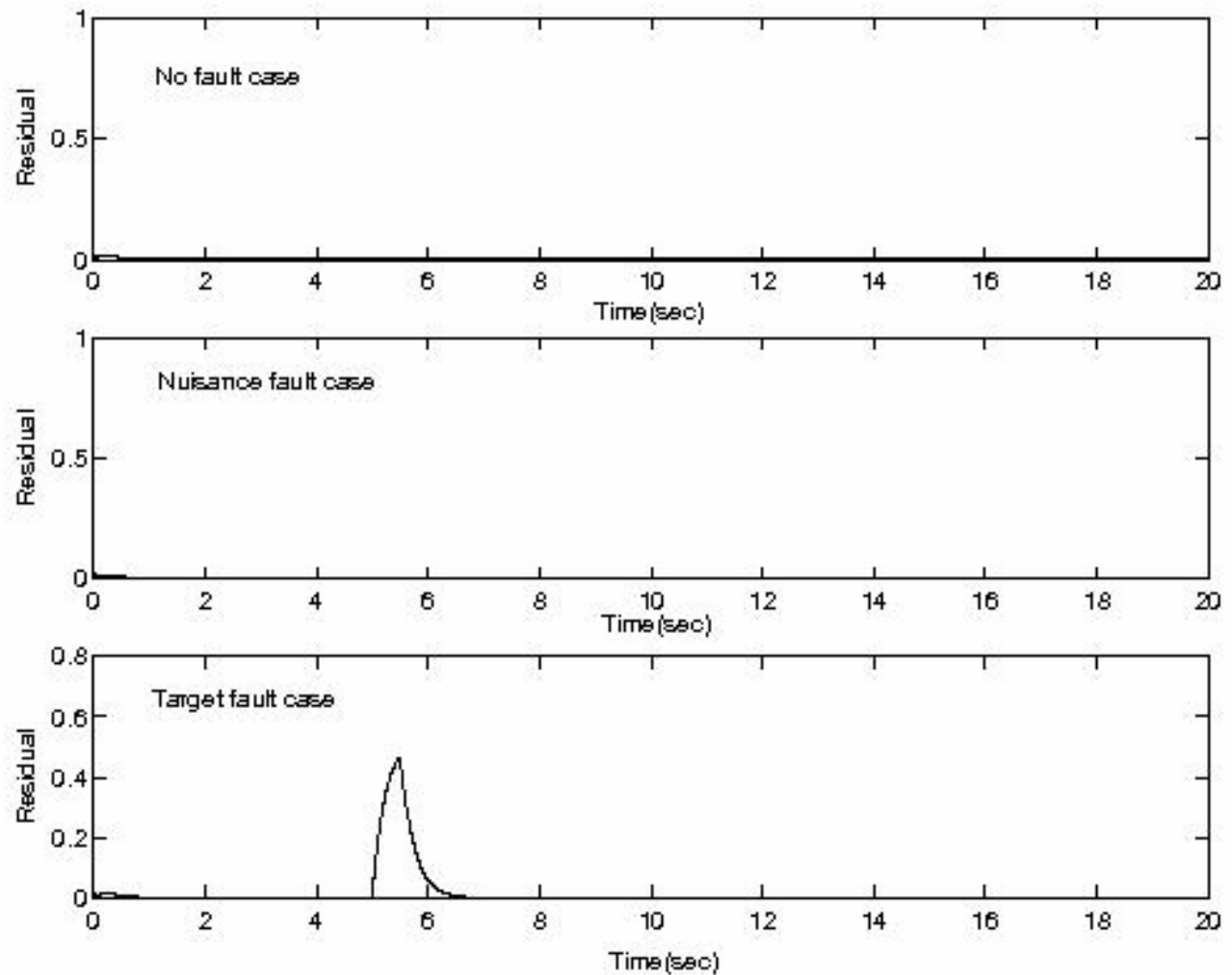
I.C.

$$x(0) = [.2, .2, .4]$$

$$\hat{x}(0) = [.5, 0, .2]$$

Fig 8. Estimation error

Fig 9. Time response of the residual at different fault cases ( $\gamma=10e-4$ )



# *Conclusions*

- The residual is primarily affected by the target fault and minimally by the nuisance faults
- New method to design a fault detection filter for nonlinear systems
- Sometimes there are problems with the solution of the Ricatti equation

# *Future Work*

- Study of the required number of approximations
- Combine fault detection with control
- Reconstruction of the fault
- Further study of the limiting case



# *General Conclusions and Recommendations*

- Present two different linearization techniques and their application to different problems in Control Theory
- The choice of the linearization technique depends on the specific problem to be solved
- Important to know ways of attacking nonlinear problems
- Further study of the previous linearization techniques and applications
- Study and comparison of the techniques

*Thank you*