Adaptive Mode Selection for Bidirectional Relay Networks – Fixed Rate Transmission

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Abstract—In this paper, we consider the problem of sum throughput maximization for bidirectional relay networks with block fading. Thereby, user 1 and user 2 exchange information only via a relay node, i.e., a direct link between both users is not present. We assume that channel state information at the transmitter (CSI) is not available and/or only one coding and modulation scheme is used at the transmitters due to complexity constraints. Thus, the nodes transmit with a fixed predefined rate regardless of the channel state information (CSI). In general, the nodes in the network can assume one of three possible states in each time slot, namely the transmit, receive, and silent state. Most of the existing protocols assume a fixed schedule for the sequence of the states of the nodes. In this paper, we abandon the restriction of having a fixed and predefined schedule and propose a new protocol which, based on the CSI at the receiver (CSIR), selects the optimal states of the nodes in each time slot such that the sum throughput is maximized. To this end, the relay has to be equipped with two buffers for storage of the information received from the two users. Numerical results show that the proposed protocol significantly outperforms the existing protocols.

I. INTRODUCTION

Relaying in cooperative communication was originally proposed as a means to extend the system coverage and to increase the throughput and reliability of wireless networks [1]. Recently, bidirectional relaying, where two users exchange information via a relay node, has attracted much interest. In particular, this simple network architecture can be used to model several practical applications such as satellite communication and cellular communication via a base station.

Several protocols have been proposed for the bidirectional relay network under the practical half-duplex constraint, i.e., a node cannot transmit and receive at the same time and in the same frequency band. The simplest protocol is the traditional two-way relaying protocol in which the bidirectional transmission is accomplished using four successive point-to-point phases: user 1-to-relay, relay-to-user 2, user 2-to-relay, and relay-to-user 1. However, this protocol suffers from a loss in spectral efficiency due to the pre-log factor of $\frac{1}{2}$ caused by the two-hop transmission architecture. To increase spectral efficiency of bidirectional relaying, the time division broadcast (TDBC) protocol was proposed in [2] which combines the relay-to-user 1 and relay-to-user 2 phases into one phase, the broadcast phase. Thereby, the relay transmits to both users simultaneously. To further enhance spectral efficiency, the multiple access broadcast (MABC) protocol was proposed in [3] where the user 1-to-relay and user 2-to-relay phases are also combined into one phase, the multiple-access phase. In the multiple-access phase, both users simultaneously transmit to the relay. A significant research effort has been dedicated to obtaining the achievable rate region of the bidirectional relay channel [2]–[5]. However, these rate regions were derived for adaptive rate transmission which requires the availability of channel state information at the transmitter (CSI) and the capability of using appropriate coding and modulation schemes such that the transmitters can adapt their transmission rates to the channel capacity. For the cases when CSIT is not available and/or only one coding and modulation scheme is used, protocols assuming adaptive rate transmission are not applicable. Instead, the transmitters have to transmit with fixed rates regardless of the channel state information (CSI) of the involved links. For fixed rate transmission, not the achievable rate region but other performance metrics such as throughput and outage probability have to be considered [6]–[8].

In general, the nodes in the network can assume one of three possible states in each time slot, namely the transmit, receive, and silent state. Among the $3^3 = 27$ possible combinations of the states of the nodes, only seven combinations are relevant in the considered bidirectional relay network due to the half-duplex constraint, see Table I. Each of these combinations is referred to as a transmission mode. The seven relevant transmission modes are given in the following: four point-to-point modes (user 1-to-relay, user 2-to-relay, relay-to-user 1, relay-to-user 2), a multiple access mode (both users to the relay), a broadcast mode (the relay to both users), and a silent mode (all nodes are silent). Previously proposed protocols utilize a fixed and predefined schedule of using a subset of the available transmission modes [1]–[8]. However, for one-way relaying, it was shown in [9] and [10] that a considerable gain is obtained with adaptive link selection where based on the CSI either the source-relay or relay-destination links are selected for transmission in each time slot. In particular, optimal link selection policies achieving the maximum throughput of one-way relay networks were derived for adaptive and fixed rate transmission in [9] and [10], respectively. For bidirectional relay networks, the optimal mode selection policy achieving the maximum sum rate was derived in [11] for adaptive rate transmission. A simpler protocol was proposed in [12] where only two point-to-point modes and the broadcast mode were available for selection. Thus, this protocol has a lower performance compared to the protocol in [11].

<table>
<thead>
<tr>
<th>Transmission Mode</th>
<th>$M_1$</th>
<th>$M_2$</th>
<th>$M_3$</th>
<th>$M_4$</th>
<th>$M_5$</th>
<th>$M_6$</th>
<th>$M_7$</th>
</tr>
</thead>
<tbody>
<tr>
<td>User 1</td>
<td>T</td>
<td>S</td>
<td>T</td>
<td>R</td>
<td>S</td>
<td>R</td>
<td></td>
</tr>
<tr>
<td>User 2</td>
<td>S</td>
<td>T</td>
<td>T</td>
<td>S</td>
<td>R</td>
<td>R</td>
<td>S</td>
</tr>
<tr>
<td>Relay</td>
<td>R</td>
<td>R</td>
<td>T</td>
<td>T</td>
<td>T</td>
<td>S</td>
<td></td>
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</table>
Motivated by the performance gains reported in [9]–[12], in this paper, we consider the problem of sum throughput maximization for the bidirectional relay network under the constraint of fixed rate transmission. In particular, we propose a protocol which is not restricted to have a fixed and predefined schedule of using the available transmission modes. Instead, based on the channel state information at the receiver (CSIR), the optimal transmission mode is selected in each time slot such that the sum throughput is maximized. To this end, the relay has to be equipped with two buffers for storage of the information received from user 1 and user 2. Considering fixed rate transmission is of both practical and theoretical interest. Specifically, simpler transmitters can be employed for fixed rate transmission than for adaptive rate transmission since only one coding and modulation scheme is needed. Moreover, the overhead required for feedback information is very low as only three bits of feedback are required in the proposed protocol to select one of the seven possible transmission modes. From a theoretical point of view, it is interesting to obtain performance limits for bidirectional relay networks under the constraint of a fixed transmission rate.

II. System Model

In this section, we describe the system model and analyze the seven possible transmission modes used to develop the proposed protocol.

A. Channel Model

We consider a simple network in which user 1 and user 2 exchange information with the help of a relay node as shown in Fig. 1. We assume that there is no direct link between the users, and thus, user 1 and user 2 communicate with each other only through the relay node. We assume that all three nodes in the network are half-duplex. Furthermore, we assume that time is divided into slots of equal length and each node transmits codewords which span one time slot. We assume that the user-to-relay and relay-to-user channels are impaired by additive white Gaussian noise (AWGN) and block fading, i.e., the channel coefficients are constant during one time slot and change from one time slot to the next. Moreover, in each time slot, the channels are assumed to be reciprocal, i.e., the user 1-to-relay and the user 2-to-relay channels are identical to the relay-to-user 1 and relay-to-user 2 channels, respectively. The channel reciprocity assumption is valid for time-division-duplex (TDD) systems where the user-to-relay and relay-to-user links utilize the same frequency band. Let $h_1(i)$ and $h_2(i)$ denote the channel fading coefficients between user 1 and the relay and between user 2 and the relay in the $i$-th time slot, respectively. Fading gains $|h_1(i)|^2$ and $|h_2(i)|^2$ are assumed to be ergodic and stationary random processes with means $\Omega_1 = E\{|h_1(i)|^2\}$ and $\Omega_2 = E\{|h_2(i)|^2\}$, respectively, where $E\{\cdot\}$ denotes expectation. Furthermore, $\gamma_1(i) = \gamma|h_1(i)|^2$ and $\gamma_2(i) = \gamma|h_2(i)|^2$ denote the instantaneous signal-to-noise ratios (SNRs) of the links between user 1 and the relay and user 2 and the relay, respectively, where $\gamma = \frac{P}{\sigma_n^2}$ is the transmit SNR of the nodes, $P$ is the transmit power of the nodes, and $\sigma_n^2$ is the noise variance at the receivers. Since the noise is AWGN, we assume that the transmitted codewords of user 1, user 2, and the relay are comprised of symbols which are Gaussian distributed random variables with variance $P$. We also assume that all nodes transmit with fixed rate $R_0$.

B. Analysis of the Transmission Modes

In the considered bidirectional relay channel, only seven transmission modes are relevant, cf. Table I. The transmission modes are denoted by $M_1, \ldots, M_7$. In order to avoid information loss, we select transmission modes $M_1, \ldots, M_6$ only if the information can be decoded successfully at the receiver(s). Otherwise, we select silent mode $M_7$. Let $B_1$ and $B_2$ denote two infinite-size buffers at the relay which store the information received from user 1 and user 2, respectively. Moreover, $Q_j(i), j \in \{1, 2\}$, denotes the amount of normalized information in bits/symbol available in buffer $B_j$ in the $i$-th time slot. Using these notations and assumptions, the transmission modes and the dynamics of the queues at the buffers are presented in the following:

$M_1$: User 1 transmits to the relay and user 2 is silent. For this mode, the relay can decode the information successfully only if $\gamma_1(i) > \gamma_{thr}$ holds, where $\gamma_{thr} = 2^{R_0} - 1$. Thereby, the relay stores the information in buffer $B_1$ and the amount of information in buffer $B_1$ increases to $Q_1(i) = Q_1(i-1) + R_0$.

$M_2$: User 2 transmits to the relay and user 1 is silent. For this mode, the relay can decode the information successfully only if $\gamma_2(i) > \gamma_{thr}$ holds. Thereby, the relay stores the information in buffer $B_2$ and the amount of information in buffer $B_2$ increases to $Q_2(i) = Q_2(i-1) + R_0$.

$M_3$: Both users 1 and 2 transmit to the relay simultaneously. For this mode, we assume that multiple access transmission is used, see [13]. The relay can decode the information from both users successfully only if $\gamma_1(i) > \gamma_{thr}$, $\gamma_2(i) > \gamma_{thr}$, and $\gamma_1(i) + \gamma_2(i) > \gamma_{sum}^{thr}$ hold, where $\gamma_{sum}^{thr} = 2^{2R_0} - 1$. Thereby, the relay stores the information received from user 1 and user 2 in buffers $B_1$ and $B_2$, respectively. Therefore, the amounts of information in buffers $B_1$ and $B_2$ increase to $Q_1(i) = Q_1(i-1) + R_0$ and $Q_2(i) = Q_2(i-1) + R_0$, respectively.

$M_4$: The relay transmits the information received from user 2 to user 1. Specifically, the relay extracts $R_0$ bits of information from buffer $B_2$, encodes it into a codeword, and transmits it to user 1. Thus, a successful transmission for this mode depends on both availability of information in buffer $B_2$ and condition of the relay-to-user 1 link. In particular, user 1 can decode $R_0$ bits of information successfully in this mode only if $\gamma_1(i) > \gamma_{thr}$ and $Q_2(i-1) \geq R_0$ hold. Thereby, the amount of information in buffer $B_2$ decreases to $Q_2(i) = Q_2(i-1) - R_0$.

$M_5$: This mode is identical to $M_4$ with user 1 and 2 switching roles. For this mode, user 2 can decode $R_0$ bits of information successfully only if $\gamma_2(i) > \gamma_{thr}$ and $Q_1(i-1) \geq R_0$ hold. Thereby, the amount of information in buffer $B_1$ decreases to $Q_1(i) = Q_1(i-1) - R_0$.

$M_6$: The relay broadcasts to both user 1 and user 2 the information received from user 2 and user 1, respectively.
Specifically, the relay extracts $R_0$ bits of information intended for user 2 from buffer $B_1$ and $R_0$ bits of information intended for user 1 from buffer $B_2$. Then, based on the scheme in [5], it constructs a superimposed codeword which contains the information of both users and broadcasts it to the users. For this mode, both users can decode $R_0$ bits of information successfully only if $\gamma_1(i) > \gamma_{thr}$, $\gamma_2(i) > \gamma_{thr}$, $Q_1(i-1) \geq R_0$, and $Q_2(i-1) \geq R_0$ hold. Thereby, the amounts of information in buffers $B_1$ and $B_2$ decrease to $Q_1(i) = Q_1(i-1) - R_0$ and $Q_2(i) = Q_2(i-1) - R_0$, respectively.

$\mathcal{M}_7$: For this mode all nodes are silent and the status of the queues at the buffers does not change.

C. Mode Selection Variables and SNR Regions

We introduce seven binary variables, $q_k(i) \in \{0,1\}$, $k = 1, \ldots, 7$, where $q_k(i)$ indicates whether or not transmission mode $\mathcal{M}_k$ is selected in the $i$-th time slot. In particular, $q_k(i) = 1$ if mode $\mathcal{M}_k$ is selected and $q_k(i) = 0$ if it is not selected in the $i$-th time slot. Furthermore, since in each time slot only one of the seven transmission modes can be selected, only one of the mode selection variables is equal to one and the others are zero, i.e., $\sum_{k=1}^{7} q_k(i) = 1$ holds.

Moreover, let $O_k(i) \in \{0,1\}$ be a binary variable specifying the decodability of information at the receivers for mode $\mathcal{M}_k$ based on CSI in the $i$-th time slot. In particular, assuming the availability of information at the transmitters, $O_k(i) = 1$ if the transmitted information can be successfully decoded at the receiver(s) for mode $\mathcal{M}_k$ and $O_k(i) = 0$ if the transmitted information cannot be successfully decoded at all receivers for mode $\mathcal{M}_k$ in the $i$-th time slot. Fig. 2 illustrates five possible regions for the instantaneous link SNRs based on the decodability of information at the receivers, i.e., $\gamma(i) = [\gamma_1(i), \gamma_2(i)] \in \mathcal{R}_l$, $l = 1, \ldots, 5$. In particular, in region $\mathcal{R}_1$, for all the modes, the receivers can decode the information. In region $\mathcal{R}_2$, only the relay cannot decode both users’ information in the multiple access mode $\mathcal{M}_3$. In regions $\mathcal{R}_3$ and $\mathcal{R}_4$, only the receivers for modes $\{\mathcal{M}_1, \mathcal{M}_4\}$ and $\{\mathcal{M}_2, \mathcal{M}_5\}$ can successfully decode the information, respectively. Finally, in region $\mathcal{R}_5$, none of the receivers for modes $\mathcal{M}_1, \ldots, \mathcal{M}_6$ can decode the information successfully. We also define $P_{\mathcal{R}_l} = \operatorname{Pr}\{\gamma(i) \in \mathcal{R}_l\}$ for future reference.

III. ADAPTIVE MODE SELECTION

In this section, we first present the problem formulation for sum throughput maximization with adaptive mode selection. Then, we propose a protocol for the optimal mode selection policy as a solution of the optimization problem. Moreover, we analyze the performance of the proposed protocol in the high SNR regime.

A. Problem Formulation

Let $\bar{R}_{12}$ and $\bar{R}_{21}$ denote the average throughputs achieved for the user 1-to-user 2 and user 2-to-user 1 transmissions, respectively. In this paper, our goal is to devise a protocol that optimally selects the transmission mode based on the instantaneous CSI in each time slot such that the sum throughput, i.e., $R_{\text{sum}} = \bar{R}_{12} + \bar{R}_{21}$, is maximized. On the other hand, the outage probability for each transmission direction is defined as the reduction of the achievable throughput compared to the maximum throughput when the receivers can always decode the information [10]. Mathematically, we can write the outage probabilities for both transmission directions as

$$F_{\text{out}}^{12} = 1 - \frac{\bar{R}_{12}}{R_{\text{max}}^{12}} \quad F_{\text{out}}^{21} = 1 - \frac{\bar{R}_{21}}{R_{\text{max}}^{21}} \quad (1)$$

where $\bar{R}_{12}^{\text{max}} = \bar{R}_{21}^{\text{max}} = R_0/2$ [10]. Therefore, by maximizing the sum throughput, the sum of the individual outage probabilities is minimized. In other words, the outage probability of the system is minimized, i.e., $F_{\text{out}}^{\text{sys}} = \frac{F_{\text{out}}^{12} + F_{\text{out}}^{21}}{2}$.

We assume that user 1 and user 2 always have enough information to send in all time slots and that the number of time slots, $N$, satisfies $N \to \infty$. Moreover, the average throughputs of the user 1-to-relay, user 2-to-relay, relay-to-user 1, and relay-to-user 2 links are denoted by $\bar{R}_{1r}$, $\bar{R}_{2r}$, $\bar{R}_{1r}$, and $\bar{R}_{r2}$, respectively, and are given by

$$\bar{R}_{1r} = \lim_{N \to \infty} \frac{1}{N} \sum_{i=1}^{N} [q_1(i)O_1(i) + q_2(i)O_2(i)] R_0 \quad (2a)$$

$$\bar{R}_{2r} = \lim_{N \to \infty} \frac{1}{N} \sum_{i=1}^{N} [q_2(i)O_2(i) + q_3(i)O_3(i)] R_0 \quad (2b)$$

$$\bar{R}_{1r} = \lim_{N \to \infty} \frac{1}{N} \sum_{i=1}^{N} [q_3(i)O_3(i) + q_6(i)O_6(i)] \min\{R_0, Q_2(i-1)\} \quad (2c)$$

$$\bar{R}_{r2} = \lim_{N \to \infty} \frac{1}{N} \sum_{i=1}^{N} [q_6(i)O_6(i) + q_8(i)O_8(i)] \min\{R_0, Q_1(i-1)\} \quad (2d)$$

Furthermore, the average information received at user 2 from user 1 is identical to the average information that user 2 receives from the relay, i.e., $\bar{R}_{12} = \bar{R}_{r2}$. Similarly, we obtain that $\bar{R}_{21} = \bar{R}_{r1}$ has to hold.

In [11, Theorem 1], we have introduced a useful condition for the queues of the buffers at the relay for adaptive rate transmission. The same condition must hold when the nodes transmit with fixed rates. In particular, the queues of the buffers $B_1$ and $B_2$ at the relay have to be at the edge of non-absorption. More precisely, for the maximum throughput, $\bar{R}_{1r} = \bar{R}_{r2}$ and $\bar{R}_{2r} = \bar{R}_{r1}$ must hold where $\bar{R}_{r1}$ and $\bar{R}_{r2}$
have to satisfy
\[ \bar{R}_{r_1} = \lim_{N \to \infty} \frac{1}{N} \sum_{i=1}^{N} [q_4(i)O_4(i) + q_6(i)O_6(i)] R_0 \] (3a)
\[ \bar{R}_{r_2} = \lim_{N \to \infty} \frac{1}{N} \sum_{i=1}^{N} [q_5(i)O_5(i) + q_6(i)O_6(i)] R_0 \] (3b)

For the proof please refer to [11]. In particular, the capacity terms for adaptive rate transmission in [11] have to be replaced by \( O_k(i)R_0 \), \( \forall i, k \) for fixed rate transmission. We note that with the above conditions, the effect of the queues at the relay becomes negligible for the maximum sum throughput, i.e., the relay always has information to transmit and thus, (2c) and (2d) simplify to (3a) and (3b), respectively.

Now, we are ready to present the considered optimization problem. The sum throughput maximization problem with adaptive mode selection is formulated as follows

\[
\begin{align*}
\text{maximize} & \quad \bar{R}_{\text{sum}} \\
\text{subject to} & \quad \bar{R}_{\gamma} = \bar{R}_{\gamma_{k+1}} \\
& \quad \sum_{k=1}^{7} q_k(i) = 1, \, \forall i \\
& \quad q_k(i) \in \{0, 1\}, \, \forall i, k
\end{align*}
\] (4)

where constraints C1 and C2 are the optimal conditions of the queues of the buffers, i.e., the queues must be at the edge of non-absorption, and constraints C3 and C4 impose the necessary restrictions on the mode selection variables.

B. Optimal Mode Selection Policy

In this subsection, we propose the optimal mode selection protocol as a solution to the optimization problem in (4). In particular, as established in Section II-C, the instantaneous SNRs of the links, \( \gamma(i) \), belong to one of the five different SNR regions shown in Fig. 2, i.e. \( \mathcal{R}_k \), \( k = 1, \ldots, 5 \). The proposed protocol selects the optimal transmission mode in each time slot based on which SNR region \( \gamma(i) \) belongs to. Moreover, the optimal mode selection policy depends on the statistics of the fading gains. Thus, we distinguish several statistical regions for the fading gains and each statistical region requires a different optimal selection policy. As we will see later, the optimal mode selection policy may require rolling a die. Therefore, we define \( X_n^M(i) \in \{1, \ldots, M\} \) as the outcome of rolling the \( n \)-th die with \( M \) faces in the \( i \)-th time slot. The probabilities of the possible outcomes of rolling the \( n \)-th die are given by \( \text{Pr}\{X_n^M(i) = m\} = p_{(m)}^{(i)} \), \( 1 \leq m \leq M \).

**Theorem 1**: For \( N \to \infty \), the optimal mode selection policy which maximizes the sum throughput of the considered half-duplex bidirectional relay network with AWGN and block fading is given by

If \( \gamma(i) \in \mathcal{R}_1 \implies \mathcal{M}_{k^*} = \begin{cases} \mathcal{M}_3, & \text{if } X_1^2(i) = 1 \\ \mathcal{M}_6, & \text{if } X_1^2(i) = 2 \end{cases} \)

If \( \gamma(i) \in \mathcal{R}_2 \implies \mathcal{M}_{k^*} = \begin{cases} \mathcal{M}_1, & \text{if } X_2^3(i) = 1 \\ \mathcal{M}_2, & \text{if } X_2^3(i) = 2 \\ \mathcal{M}_6, & \text{if } X_2^3(i) = 3 \end{cases} \)

If \( \gamma(i) \in \mathcal{R}_3 \implies \mathcal{M}_{k^*} = \begin{cases} \mathcal{M}_1, & \text{if } X_3^3(i) = 1 \\ \mathcal{M}_5, & \text{if } X_3^3(i) = 2 \\ \mathcal{M}_7, & \text{if } X_3^3(i) = 3 \end{cases} \)

If \( \gamma(i) \in \mathcal{R}_4 \implies \mathcal{M}_{k^*} = \begin{cases} \mathcal{M}_2, & \text{if } X_4^3(i) = 1 \\ \mathcal{M}_5, & \text{if } X_4^3(i) = 2 \\ \mathcal{M}_7, & \text{if } X_4^3(i) = 3 \end{cases} \)

If \( \gamma(i) \in \mathcal{R}_5 \implies \mathcal{M}_{k^*} = \mathcal{M}_7 \)

where the probabilities for the outcomes of rolling the dice depend on the statistics of the channel gains and are given in Table II. With this protocol, the maximum sum throughput and minimum system outage probability are given by

\[
\bar{R}_{\text{sum}} = \frac{(P_{R_1} + P_{R_2} + P_{\min})R_0}{2P_{R_1} + P_{R_2} + P_{\min} + P_{R_3} + P_{\min}R_0}, \quad \text{if } \frac{P_{R_2} - P_{R_1}}{P_{\min}} < 2\frac{P_{\max}}{P_{\min}} - 1
\] (6)

\[
P_{\text{out}} = \frac{P_{R_5} + P_{\max}}{\frac{1}{2} - \frac{2}{3} (P_{R_1} - P_{R_2})}, \quad \text{otherwise}
\] (7)

where \( P_{\max} = \max\{P_{R_3}, P_{R_4}\} \) and \( P_{\min} = \min\{P_{R_3}, P_{R_4}\} \).

**Proof**: Please refer to Appendix A.

We note that the die probabilities in the optimal mode selection policy in Theorem 1 depend only on the long term statistics of the channel gains. Hence, they can be obtained offline and used as long as the channel statistics remain unchanged. Moreover, we give the probabilities of \( M-1 \) faces for a die with \( M \) faces in Table II due to space constraints. The last probability is obtained as \( p_{(m)}^{(i)} = 1 - \sum_{m=1}^{M-1} p_{(m)}^{(i)} \).

**Remark 1**: We note that for the case \( P_{R_3} \leq P_{R_4} \) and \( \frac{P_{R_2} - P_{R_1}}{P_{\min}} \leq 1 \), we have a degree of freedom in choosing \( p_1^{(1)} \) and \( p_1^{(2)} \) as only \( p_4^{(1)} + p_4^{(2)} = \frac{P_{R_3}}{P_{R_4}} \) has to hold, cf. Table II. A similar degree of freedom exists for \( p_1^{(3)} \) and \( p_1^{(4)} \) for the case \( P_{R_3} \geq P_{R_4} \) and \( \frac{P_{R_2} - P_{R_1}}{P_{\min}} \leq 1 \) as only \( p_3^{(1)} + p_3^{(2)} = \frac{P_{R_3}}{P_{R_4}} \) has to hold.

**Remark 2**: We observe that the optimal mode selection policy adopts mode \( \mathcal{M}_2 \) in instantaneous SNR regions \( \mathcal{R}_3 \) and \( \mathcal{R}_4 \) for some channel statistics, cf. Theorem 1 and Table II. Hence, although one of the links could support the transmission in this case, the optimal mode selection policy forces all nodes to be silent. The reason for this is that one of the channels is statistically weaker than the other one and the maximum sum throughput is limited by the weaker channel. Therefore, if the transmission modes associated with the stronger channel are always selected, the weaker channel cannot convey the information and consequently constraints C1 and C2 in (4) are violated.

**Remark 3**: We assume that the relay is responsible for performing the optimal mode selection using the protocol in Theorem 1. In particular, in the beginning of each time slot, the users send pilots to the relay. The relay has to determine the instantaneous SNR region \( \mathcal{R}_i, i = 1, \ldots, 5 \), to select the optimal transmission mode according to Theorem 1. Then, the relay broadcasts the optimal transmission mode to the users using three bits of feedback and transmission begins. We note that the relay has to determine only the instantaneous SNR region \( \mathcal{R}_i \), which is in general a less strict requirement compared to determining the exact values of the instantaneous SNR.
C. High SNR Analysis

For Rayleigh fading, the outage probability of the protocol in Theorem 1 in the high SNR regime, i.e., \( \gamma \to \infty \), is given by

\[
\begin{align*}
R_{\text{sum}} &= R_0 \quad \text{and} \quad F_{\text{sys}}^{\text{out}} = P_{\text{max}}. 
\end{align*}
\]

For Rayleigh fading, the outage probability of the system simplifies to

\[
F_{\text{sys}}^{\text{out}} = \frac{\bar{\gamma}^{\Omega}}{\gamma_{\text{min}}} \cdot \frac{1}{\bar{\gamma}} + o\left(\frac{1}{\gamma}\right) \tag{9}
\]

where \( \Omega_{\min} = \min\{\Omega_1, \Omega_2\} \).

Proof: For the high SNR regime, i.e., \( \gamma \to \infty \), we obtain \( P_{R_1} \to 1 \) and \( P_{R_2} \to 0 \) which leads to statistical region \( \frac{P_{R_2} - P_{R_1}}{P_{\text{min}}} < 0 < \frac{2P_{\text{max}}}{P_{\text{min}}}-1 \). Therefore, from (6) and (7) and knowing that \( \frac{P_{R_2} + P_{\text{min}}}{P_{R_1}} \to 0 \) and \( \frac{P_{R_2} + P_{\text{sys}}}{P_{\text{max}}} \to 0 \) hold as \( \gamma \to \infty \), we obtain \( R_{\text{sum}} = R_0 \) and \( F_{\text{sys}}^{\text{out}} = P_{\text{max}} \).

For Rayleigh fading, the probability density functions (pdfs) of \( \gamma_1(i) \) and \( \gamma_2(i) \) are given by

\[
\begin{align*}
f_{\gamma_1}(\gamma_1) &= \frac{1}{\bar{\gamma}_{\text{min}}} e^{-\frac{\gamma_1}{\bar{\gamma}_{\text{min}}}} \quad \text{and} \quad f_{\gamma_2}(\gamma_2) = \frac{1}{\bar{\gamma}_{\text{max}}} e^{-\frac{\gamma_2}{\bar{\gamma}_{\text{max}}}}.
\end{align*}
\]

Using the Taylor series \( e^x = 1 + x + o(x) \) for \( x \to 0 \), we obtain \( F_{\text{sys}}^{\text{out}} \) in (9). This completes the proof.

IV. Numerical Results

In this section, we numerically evaluate the performance of the proposed protocol for the considered bidirectional relay network in Rayleigh fading. Moreover, due to space constraints, all presented results are obtained for \( \Omega_1 = \Omega_2 = 1 \) and \( R_0 = 1 \) for all SNR, \( \gamma \).

We adopt the traditional two-way, TDBC, and MABC protocols as benchmark schemes [2], [3]. In order to obtain a fair comparison with respect to the delay, we modify the benchmark schemes such that they also exploit the buffering capability. In this case, although the benchmark protocols have a fixed and predetermined schedule of transmission, the users are allowed to transmit to the relay for a fraction of \( N \) time slots consecutively and the relay stores the information in its infinite-size buffers. Then, the relay forwards the information to the users in the remaining time slots. The fraction of \( N \) time slots allocated to each transmission mode in the benchmark schemes is optimized for maximization of the sum throughput.

In Fig. 3, we illustrate the maximum achievable sum
throughput, $\bar{R}_{\text{sum}}$, versus the transmit SNR of the nodes, $\gamma$. We observe that the maximum achievable sum throughput is saturated for all protocols at high SNR, i.e., the multiplexing gain is zero because transmit rate $R_0$ is fixed for all SNR. However, the proposed protocol and the MABC protocol achieve a maximum sum throughput of $R_0$ while the TDBC protocol and the traditional two-way protocol achieve maximum sum throughputs of $\frac{2R_1}{\gamma}$ and $\frac{2R_2}{\gamma}$, respectively. Moreover, the proposed protocol has a superior performance compared to the benchmark schemes for all SNR.

In Fig. 4, we depict the outage probability of the system versus the transmit SNR of the nodes. We note that the protocol in Theorem 1 was derived such that $\bar{R}_{\text{sum}}$ is maximized or equivalently $F^\text{out}_{\text{sum}}$ is minimized. However, as stated in Remark 1, the per-user throughputs and outage probabilities are not fixed due to the available degree of freedom in choosing the die probabilities. In Fig. 4, we also show the per-user outage probabilities $F^\text{out}_{12}$ and $F^\text{out}_{21}$ of the proposed protocol for the case where the die probabilities are chosen such that rate $R_{12}$ has its maximum value. We observe from Fig. 4 that the proposed protocol outperforms the benchmark schemes significantly. In particular, we obtain around 4 dB SNR gain compared to the best benchmark scheme, the MABC protocol. We note that the per-user outage probabilities can change between the curves of $F^\text{out}_{12}$ and $F^\text{out}_{21}$ depicted in Fig. 4 depending on how the die probabilities are chosen.

V. Conclusion

We derived a protocol which maximizes the sum throughput of bidirectional relay networks with block fading when all nodes transmit with a fixed rate. The proposed protocol selects the optimal transmission mode in each time slot based on CSIR knowledge. For this to be possible, the relay has to be equipped with two buffers for storage of the information received from the users. We also obtain the diversity-multiplexing trade-off of the proposed protocol. Our numerical results showed that the proposed protocol outperforms the existing protocols significantly in terms of achievable sum throughput and system outage probability.

APPENDIX A

PROOF OF THEOREM 1

Because of space constraints, a detailed proof of Theorem 1 is provided in [14, Appendix A] which is an extended version of the submitted paper. Herein, we only highlight the main steps in obtaining the protocol. We first relax the binary condition for $q_k(i) \in \{0, 1\}$ to $0 \leq q_k(i) \leq 1$. Moreover, since the relaxed optimization problem is a linear programming problem, an optimal solution is achieved at the boundary of the constraints, i.e., by binary $q_k(i)$. Therefore, the binary relaxation does not affect the maximum achievable sum throughput, please see [14, Appendix A] for detailed proof. In the following, we investigate the Karush-Kuhn-Tucker (KKT) conditions for the relaxed optimization problem. We note that the relaxed optimization problem is a linear program and thus, the KKT conditions are both necessary and sufficient conditions for optimality.

To simplify the usage of the KKT conditions, we change the maximization of $R_{12} + R_{21}$ to the minimization of $-(R_{1r} + R_{2r})$ since due to constraints C1 and C2 in (4), $R_{12} = R_{21}$ and $R_{1r} = R_{2r}$ must hold. Moreover, without loss of generality, we write all inequalities in the form $f(x) \leq 0$. The Lagrangian function for the relaxed optimization problem is given by

$$\mathcal{L} = -(R_{1r} + R_{2r}) + \mu_1(R_{1r} - R_{12}) + \mu_2(R_{2r} - R_{21})$$
$$+ \sum_{i=1}^{N} \lambda(i) \left( \sum_{k=1}^{7} q_k(i) - 1 \right)$$
$$+ \sum_{i=1}^{N} \sum_{k=1}^{7} \alpha_k(i) (q_k(i) - 1) - \sum_{i=1}^{N} \sum_{k=1}^{7} \beta_k(i) q_k(i)$$

where $\mu_1, \mu_2$, and $\lambda(i)$ are the Lagrange multipliers corre-
Corresponding to constraints C1, C2, and C3 in (4), respectively. Furthermore, $\alpha_k(i)$ and $\beta_k(i)$ are the Lagrange multipliers corresponding to the upper and lower limits of constraint $0 \leq q_k(i) \leq 1$, respectively.

In order to determine the optimal selection policy, $q_k^*(i)$, we have to calculate the derivatives of the Lagrangian function in (11) with respect to $q_k(i)$. This leads to

$$\frac{\partial L}{\partial q_1(i)} = -\frac{1}{N}(1-\mu_1)O_1(i)R_0 + \lambda(i) + \alpha_1(i) - \beta_1(i) \equiv 0$$

$$\frac{\partial L}{\partial q_2(i)} = -\frac{1}{N}(1-\mu_2)O_2(i)R_0 + \lambda(i) + \alpha_2(i) - \beta_2(i) \equiv 0$$

$$\frac{\partial L}{\partial q_3(i)} = -\frac{1}{N}(2-\mu_1-\mu_2)O_3(i)R_0 + \lambda(i) + \alpha_3(i) - \beta_3(i) \equiv 0$$

$$\frac{\partial L}{\partial q_4(i)} = -\frac{1}{N}\mu_2O_4(i)R_0 + \lambda(i) + \alpha_4(i) - \beta_4(i) \equiv 0$$

$$\frac{\partial L}{\partial q_5(i)} = -\frac{1}{N}O_5(i)R_0 + \lambda(i) + \alpha_5(i) - \beta_5(i) \equiv 0$$

$$\frac{\partial L}{\partial q_6(i)} = -\frac{1}{N}(\mu_1+\mu_2)O_6(i)R_0 + \lambda(i) + \alpha_6(i) - \beta_6(i) \equiv 0$$

$$\frac{\partial L}{\partial q_7(i)} = \lambda(i) + \alpha_7(i) - \beta_7(i) = 0.$$  

Without loss of generality, we first obtain the necessary condition for $q_1^*(i) = 1$ and then generalize the result to $q_k^*(i) = 1, k = 2, \ldots, 7$. If $q_k^*(i) = 1$, from constraint C3 in (4), the other selection variables are zero, i.e., $q_k^*(i) = 0, k = 2, \ldots, 7$. Furthermore, from the complementary slackness condition, we obtain that if an inequality is inactive, the respective Lagrange multiplier must be zero, i.e., $\alpha_k(i) = 0, k = 2, \ldots, 7$ and $\beta_1(i) = 0$ have to hold. By substituting these values into (12), we obtain

$$\lambda(i) + \alpha_1(i) = (1-\mu_1)O_1(i)R_0 \triangleq \Lambda_1(i)$$

$$\lambda(i) - \beta_1(i) = (1-\mu_2)O_2(i)R_0 \triangleq \Lambda_2(i)$$

$$\lambda(i) - \beta_3(i) = (2-\mu_1-\mu_2)O_3(i)R_0 \triangleq \Lambda_3(i)$$

$$\lambda(i) - \beta_4(i) = \mu_2O_4(i)R_0 \triangleq \Lambda_4(i)$$

$$\lambda(i) - \beta_5(i) = \mu_1O_5(i)R_0 \triangleq \Lambda_5(i)$$

$$\lambda(i) - \beta_6(i) = (\mu_1+\mu_2)O_6(i)R_0 \triangleq \Lambda_6(i)$$

$$\lambda(i) - \beta_7(i) = 0 \triangleq \Lambda_7(i),$$

where $\Lambda_k(i)$ is referred to as selection metric. By subtracting (13b)-(13g) from (13a), we obtain

$$\Lambda_1(i) - \Lambda_k(i) = \alpha_1(i) + \beta_k(i), \quad k = 2, \ldots, 7. \quad (14)$$

Moreover, the dual feasibility conditions for Lagrange multipliers imply $\alpha_k(i), \beta_k(i) \geq 0$. By inserting $\alpha_k(i), \beta_k(i) \geq 0$ in (14), we obtain a necessary condition for $q_1^*(i) = 1$ as

$$\Lambda_1(i) \geq \max \{\Lambda_2(i), \Lambda_3(i), \Lambda_4(i), \Lambda_5(i), \Lambda_6(i), \Lambda_7(i)\}.$$  

Repeating the same procedure for $q_k^*(i) = 1, k = 2, \ldots, 7$, we obtain a necessary condition for selecting transmission mode $M_k^*$ in the $i$-th time slot as

$$\Lambda_k^*(i) \geq \max_{k \neq k^*} \{\Lambda_k(i)\}. \quad (16)$$

Hence, we can conclude that the transmission modes with the largest selection metrics are the candidates for optimal mode selection.

Now, in order to obtain the optimal transmission mode, we need to answer the following questions: 1) what are the optimal values of selection weights $\mu_1$ and $\mu_2$ in order to calculate the selection metrics, and 2) given $\mu_1$ and $\mu_2$, how should we select the optimal mode from the modes with identical selection metric values. In [14, Appendix A], we proved that $0 \leq \mu_1 + \mu_2 \leq 1, 1 - 2\mu_1 \leq \mu_2 \leq 1, \quad \text{and} \quad 1 - 2\mu_2 \leq \mu_1 \leq 1$ have to hold to fulfill constraints C1 and C2 in (4). Moreover, the exact values of $\mu_1$ and $\mu_2$ lead to the statistical regions that are given in Table II. We also show in [14, Appendix A] that for the modes with identical selection metric values, different selection policies that satisfy constraints C1 and C2 in (4) may lead to the same sum throughput. Therefore, for simplicity, we adopt a probabilistic approach via rolling dice where the die probabilities are chosen to satisfy C1 and C2 in (4) and are given in Table II. This completes the proof.

REFERENCES


